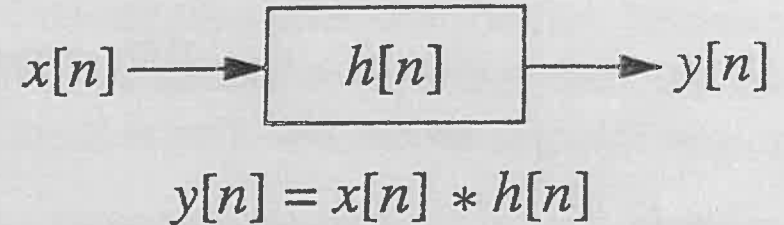


Demystifying DT Convolution Sum.

## 11.1 Graphical Flip/Shift Method

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

4. add  
2. shift  
1. flip  
3. multiply



$h[n-k]$  is viewed as function of  $k$  with fixed  $n$ .

$h[n-k] = h[-(k-n)]$  is a time-reversed and shifted version of impulse response  $h[k]$ .

Looking at the equation for convolution, we can arrive at the following step-by-step procedure:

1. Choose one signal to be  $x[n]$ , the other is then  $h[n]$ ; draw them both on the  $k$  axis.
2. *FLIP*  $h[k]$  about  $k=0$ .
3. *SHIFT* flipped version of  $h$  to the right by  $n$ .
4. *MULTIPLY*  $x[k]$  by the flipped/shifted version of  $h[k]$  and *ADD* across all values of
5. The summation in step 4 gives you  $y[n]$  for only one value of  $n$ .
6. Repeat steps 3-5 for all possible values of  $n$ .

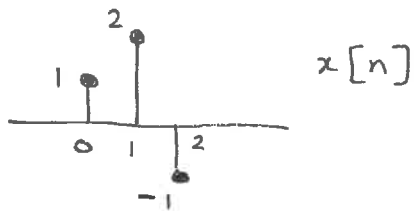
# Convolution of finite length DT Sequences

## Checks

$$y[n] = x[n] * h[n]$$

- ① Length ( $y[n]$ ) = Length ( $x[n]$ ) + Length ( $h[n]$ ) - 1
- ②  $y[n]$  will begin at Start ( $x[n]$ ) + Start ( $h[n]$ )
- ③  $y[n]$  will finish at End ( $x[n]$ ) + End ( $h[n]$ )

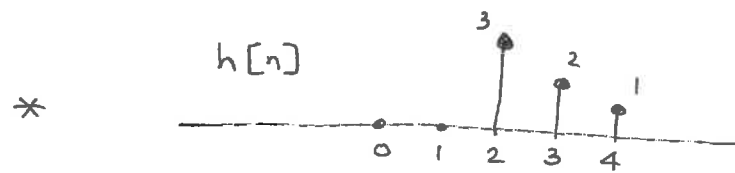
## Example



$$\text{Length}(x[n]) = 3$$

$$\text{Start}(x[n]) = 0$$

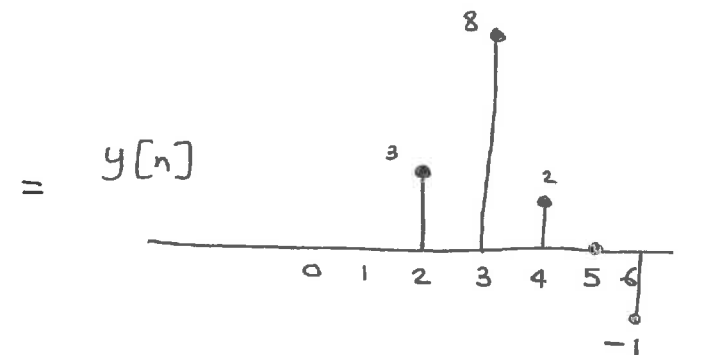
$$\text{End}(x[n]) = 2$$



$$\text{Length}(h[n]) = 3$$

$$\text{Start}(h[n]) = 2$$

$$\text{End}(h[n]) = 4$$



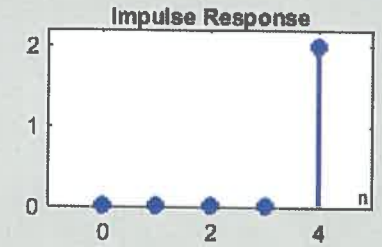
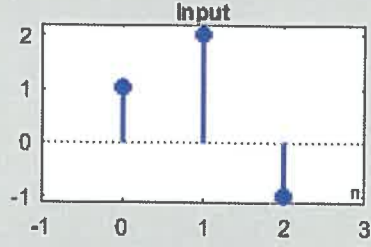
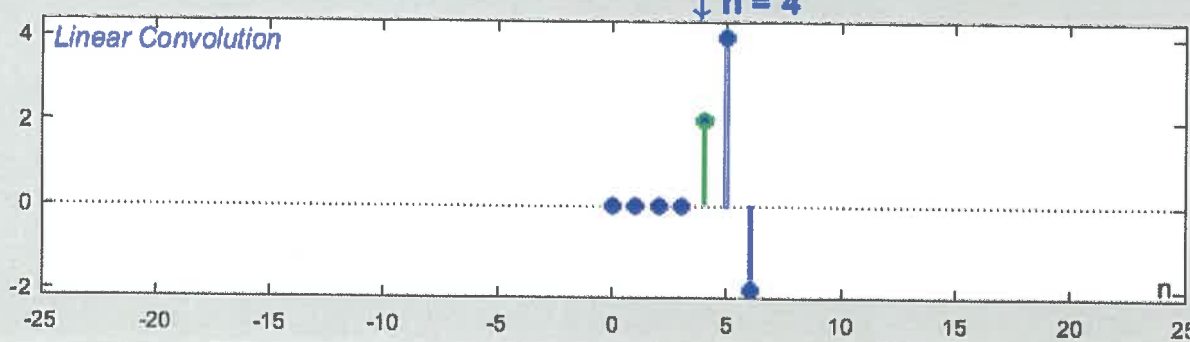
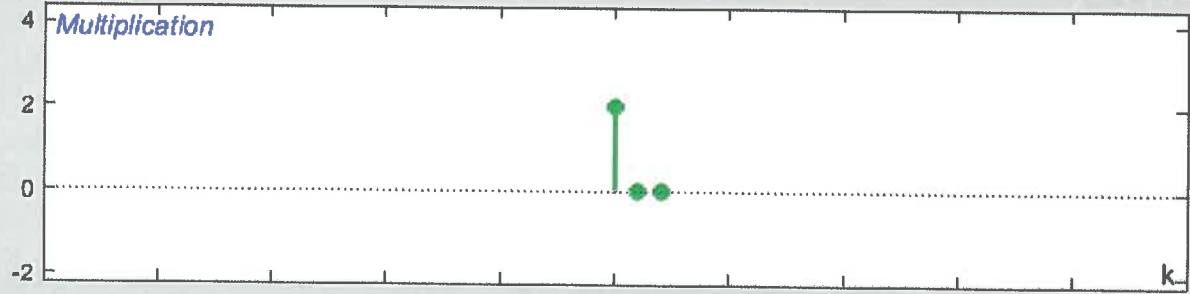
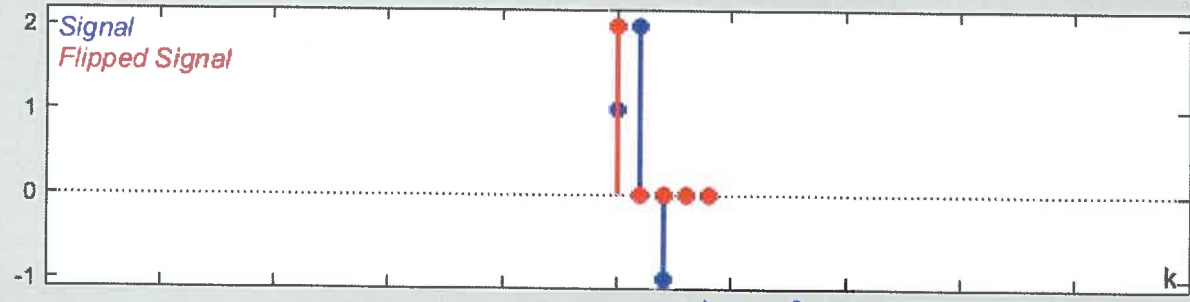
$$\begin{aligned} \text{Length}(y[n]) &= 3 + 3 - 1 \\ &= 5 \end{aligned}$$

$$\text{Start}(y[n]) = 0 + 2 = 2$$

$$\text{End}(y[n]) = 2 + 4 = 6$$

# Convolution with impulses

$x[n] = \delta[n] + 2\delta[n-1] - \delta[n-2]$      $h[n] = 2\delta[n-4]$



Get x[n]

Get h[n]

Flip x[n]

Flip h[n]

Signal Axis:

- = x[k]
- = h[n - k]

Multiplication Axis:

- = x[k]h[n - k]

Convolution Axis:

$$y[n] = \sum x[k]h[n - k]$$

Close

Help

$y[n] = 2\delta[n-4] + 4\delta[n-5] - 2\delta[n-6]$

## Convolution with Impulses

of a signal  
Convolution with a shifted impulse simply shifts the signal to the location of the impulse.

$$\delta[n] * \delta[n-4] = \delta[n-4]$$

$$\delta[n-1] * \delta[n-4] = \delta[n-4-1] = \delta[n-5]$$

↑  
replace this n with n-4

$$\delta[n-2] * \delta[n-4] = \delta[n-6]$$

$$\begin{aligned} (\delta[n] + 2\delta[n-1] - \delta[n-2]) * 2\delta[n-4] &= 2\delta[n-4] + 4\delta[n-4-1] - 2\delta[n-4-2] \\ &= 2\delta[n-4] + 4\delta[n-5] - 2\delta[n-6] \end{aligned}$$

↑  
Concept

Example

$$\begin{aligned} &u[n] * \{\delta[n] - \delta[n-1]\} \\ &= u[n] * \delta[n] - u[n] * \delta[n-1] \\ &= u[n] - u[n-1] \end{aligned}$$

↑  
replace n by n-1

Example 1

Convolution of infinite length<sup>DT</sup> sequences

$$y[n] = \underbrace{\left(\frac{1}{2}\right)^n u[n-2]}_{x[n]} * \underbrace{u[n]}_{h[n]}$$

For  $n < 2$ , there is no overlap between the non-zero portions of  $x[k]$  and  $h[n-k]$ . Consequently

$$y[n] = 0.$$

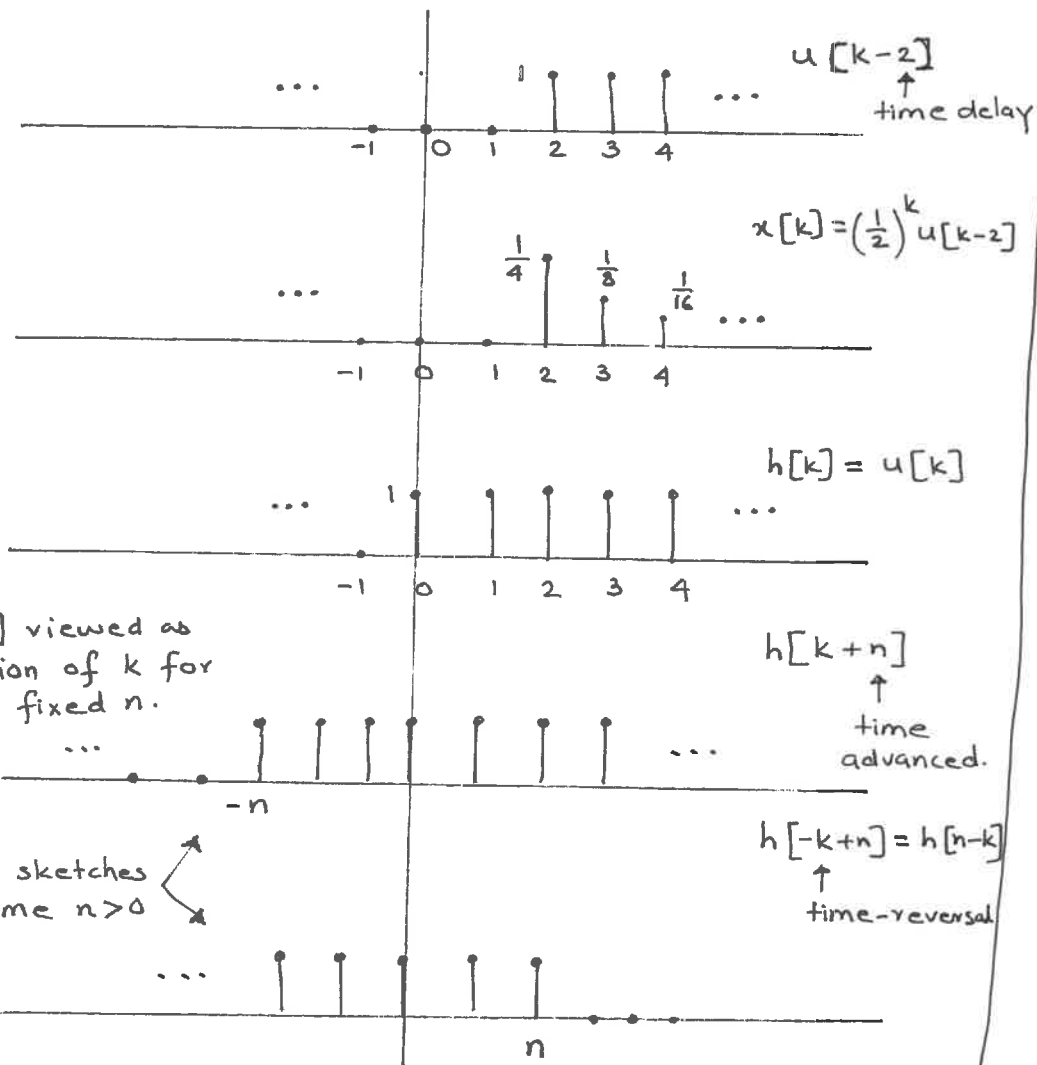
For  $n \geq 2$

$$\begin{aligned} y[n] &= \sum_{k=-\infty}^{\infty} x[k] h[n-k] \\ &= \sum_{k=2}^n \left(\frac{1}{2}\right)^k \end{aligned}$$

Using formula from sheet

$$\begin{aligned} y[n] &= \frac{\left(\frac{1}{2}\right)^2 - \left(\frac{1}{2}\right)^{n+1}}{1 - \frac{1}{2}} = \frac{\left(\frac{1}{2}\right)^2}{\left(\frac{1}{2}\right)} - \frac{\left(\frac{1}{2}\right)^{n+1}}{\left(\frac{1}{2}\right)} \\ &= \frac{1}{2} - \left(\frac{1}{2}\right)^{-1} \left(\frac{1}{2}\right)^{n+1} = \left(\frac{1}{2}\right) - \left(\frac{1}{2}\right)^n. \end{aligned}$$

$$y[n] = \begin{cases} 0 & n < 2 \\ \frac{1}{2} - \left(\frac{1}{2}\right)^n & n \geq 2 \end{cases}$$

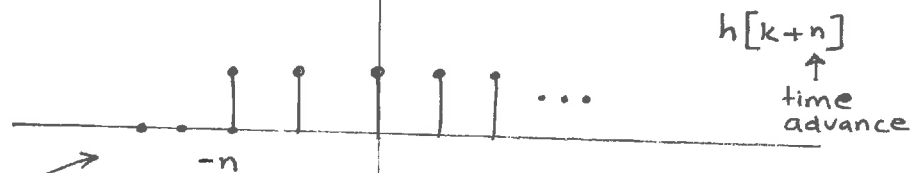
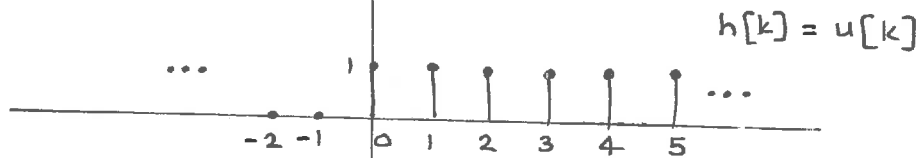
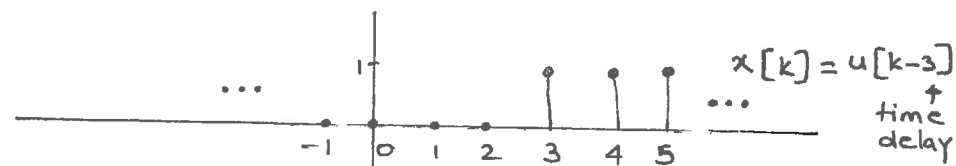


## Example 2

$$y[n] = x[n] * h[n]$$

$$x[n] = u[n-3]$$

$$h[n] = u[n]$$



$$h[-k+n] = h[n-k]$$

↑  
time reversal.



Sketches assume  $n > 0$

For  $n < 3$ , there is no overlap between non-zero samples of  $x[k]$  and  $h[n-k]$ . Consequently

$$y[n] = 0$$

For  $n \geq 3$

$$y[n] = \sum_{k=-\infty}^{\infty} x[k] h[n-k]$$

$$= \sum_{k=3}^n (1)^k \quad \left\{ \begin{array}{l} \text{Using fact that} \\ 1 = 1^k \text{ for } k > 0 \end{array} \right.$$

$$= n - 3 + 1 = n - 2.$$

Alternatively,

$$y[3] = 1, \quad n=3$$

$$y[4] = 2, \quad n=4$$

$$y[5] = 3, \quad n=5$$

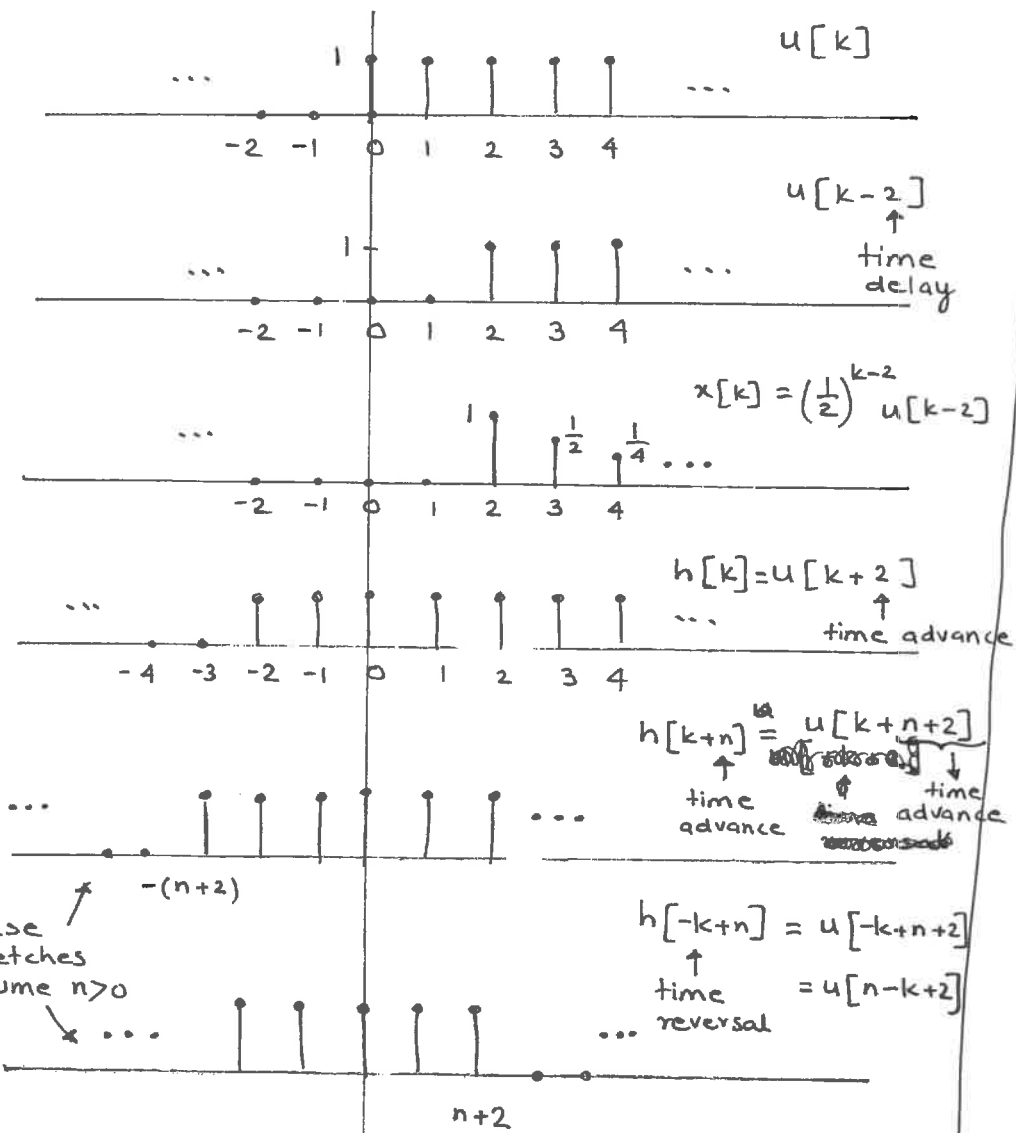
$$\therefore y[n] = n - 2, \quad n \geq 3$$

$$\text{Hence } y[n] = \begin{cases} 0 & n < 3 \\ n - 2 & n \geq 3 \end{cases}$$

Example 3

$$x[n] = \left(\frac{1}{2}\right)^{n-2} u[n-2]$$

$$h[n] = u[n+2]$$



$$\text{Hence } y[n] = \begin{cases} 0 & n < 0 \\ 2 \cdot 2^{-n} & n \geq 0 \end{cases}$$

For  $n+2 < 2$

$\Rightarrow n < 0$ , there is no overlap between the non-zero portions of  $x[k]$  and  $h[n-k]$ . Consequently

$$y[n] = 0$$

For  $n+2 \geq 2$

$\Rightarrow n > 0$

$$\begin{aligned} y[n] &= \sum_{k=-\infty}^{\infty} x[k] h[n-k] \\ &= \sum_{k=2}^{n+2} \left(\frac{1}{2}\right)^{k-2} \end{aligned}$$

Changing variable of summation from  $k$  to  $r = k-2$

$$y[n] = \sum_{r=0}^n \left(\frac{1}{2}\right)^r$$

Using formula sheet

$$\begin{aligned} y[n] &= \frac{1 - \left(\frac{1}{2}\right)^{n+1}}{\frac{1}{2}} = 2 \left(1 - \left(\frac{1}{2}\right)^{n+1}\right) \\ &= 2 \left(1 - 2^{-n-1}\right) \\ &= 2 - 2^{-n} \end{aligned}$$



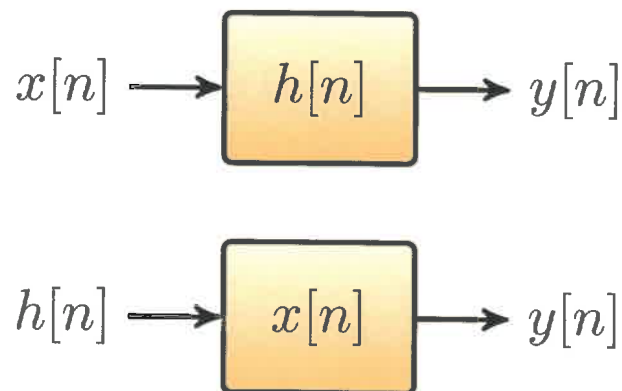
## Convolution Properties

Study from slides 177 - 182.



Previously we noted the **Commutative Property**:

$$y[n] = x[n] \star h[n] = h[n] \star x[n]$$



# DT System Properties – Commutative Property

- This follows from (change variables to  $\ell = n - k$ )

$$\begin{aligned}y[n] &= \sum_{k=-\infty}^{\infty} x[k] h[n - k] \\ &= \sum_{\ell=-\infty}^{\infty} x[n - \ell] h[\ell] = \sum_{\ell=-\infty}^{\infty} h[\ell] x[n - \ell]\end{aligned}$$

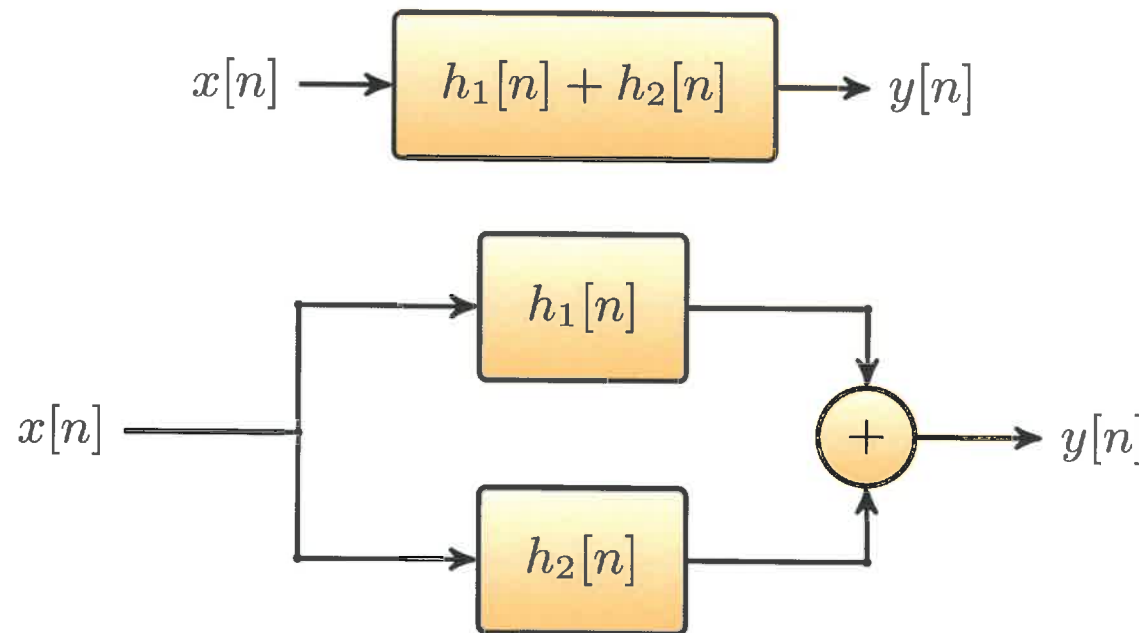
- Alternatively, if we have two polynomials say  $p(x)$  and  $q(x)$  then  $p(x)q(x) = q(x)p(x)$ . Polynomial multiplication is commutative. Convolution is commutative.

# DT System Properties – Distributive Property



Consider an input signal  $x[n]$  and two DT LTI Systems  $h_1[n]$  and  $h_2[n]$ , in parallel, then we have the **Distributive Property**:

$$x[n] \star (h_1[n] + h_2[n]) = x[n] \star h_1[n] + x[n] \star h_2[n]$$



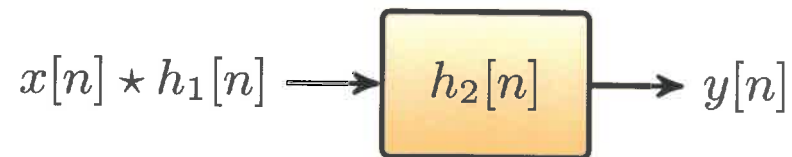
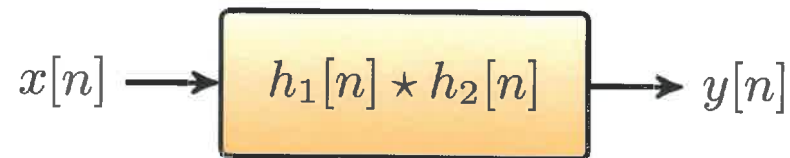
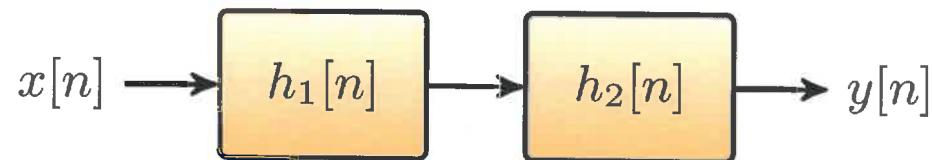
- This implies that we can combine two DT LTI systems in parallel into a single equivalent DT LTI system (by **adding** the pulse responses).

# DT System Properties – Associative Property



Consider an input signal  $x[n]$  to two DT LTI Systems  $h_1[n]$  and  $h_2[n]$ , in **cascade**, then we have the **Associative Property**:

$$x[n] \star (h_1[n] \star h_2[n]) = (x[n] \star h_1[n]) \star h_2[n]$$



- This implies that we can combine two DT LTI systems in series into a single equivalent DT LTI system (by **convolving** the pulse responses).